

SCT2220

REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Rev 1.0 : Released to Market

Rev 1.3: Update Iq in page1 and correct typo in EC table

DEVICE ORDER INFORMATION

PART NUMBER

SCT2220

ELECTRICAL CHARACTERISTICS

V_{IN}=12V, T_J=-40°C~125°C, typical values are tested under 25°C.

SYMBOL	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT
Power Supply and Output						
V _{IN}	Operating input voltage		4.2		17	V
V _{IN_UVLO}	Input UVLO Hysteresis	V _{IN} rising		3.9 300	4.15	V mV
I _{SD}	Shutdown current	EN=0, No load, V _{IN} =12V		1.5	5	uA
I _Q	Quiescent current	EN=2V, No load, No switching. V _{IN} =12V. BST-SW=5V		135		uA
Enable, Soft Start and Working Modes						
V _{EN_H}	Enable high threshold			1.2	1.25	V
V _{EN_L}	Enable low threshold		1.03	1.1		V
I _{EN}	Enable pin input current	EN=1V	1	1.5	2	uA
		EN=1.5V		6.8		uA
Power MOSFETs						
R _{DS(on)_H}	High side FET on-resistance			90		
R _{DS(on)_L}	Low side FET on-resistance			65		
Feedback and Error Amplifier						
V _{FB}	Feedback Voltage		0.78	0.8	0.82	V
Current Limit						
I _{LIM_LSD}	LSD valley current limit		2	2.8	3.6	A
Switching Frequency						
F _{SW}	Switching frequency	V _{IN} =12V, V _{OUT} =5V		750		kHz
t _{ON_MIN}	Minimum on-time*			90		ns
t _{OFF_MIN}	Minimum off-time			220		ns
Soft Start Time						
t _{SS}	Internal soft-start time			2.5		ms
Protection						
T _{SD}	Thermal shutdown threshold*	T _J rising		160		°C
	Hysteresis			20		

*Derived from bench characterization

TYPICAL CHARACTERISTICS

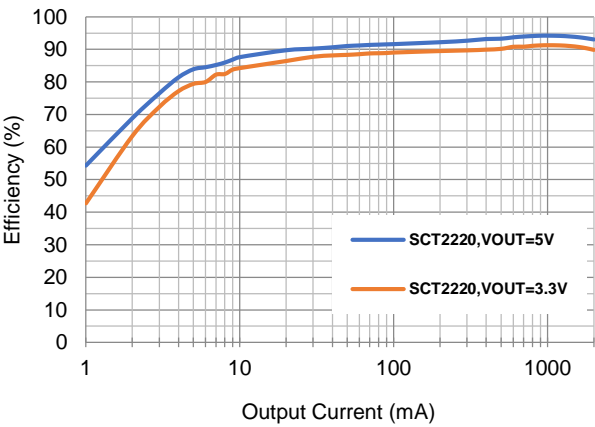


Figure 1. SCT2220 Efficiency, Vin=12V

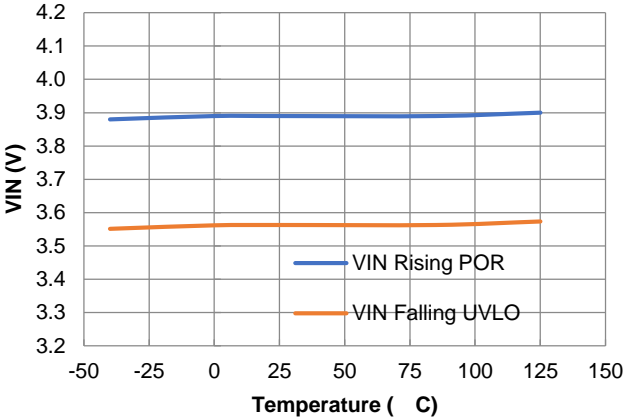


Figure 2. UVLO Vs. Temperature

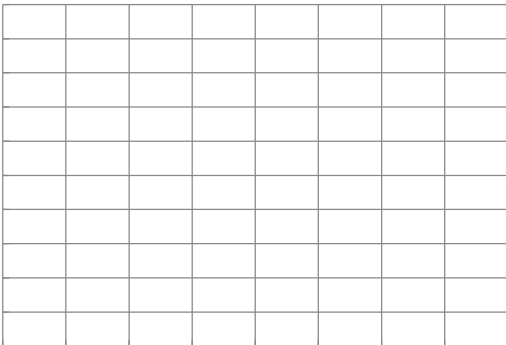


Figure 3. Line Regulation

Figure 4. Load Regulation

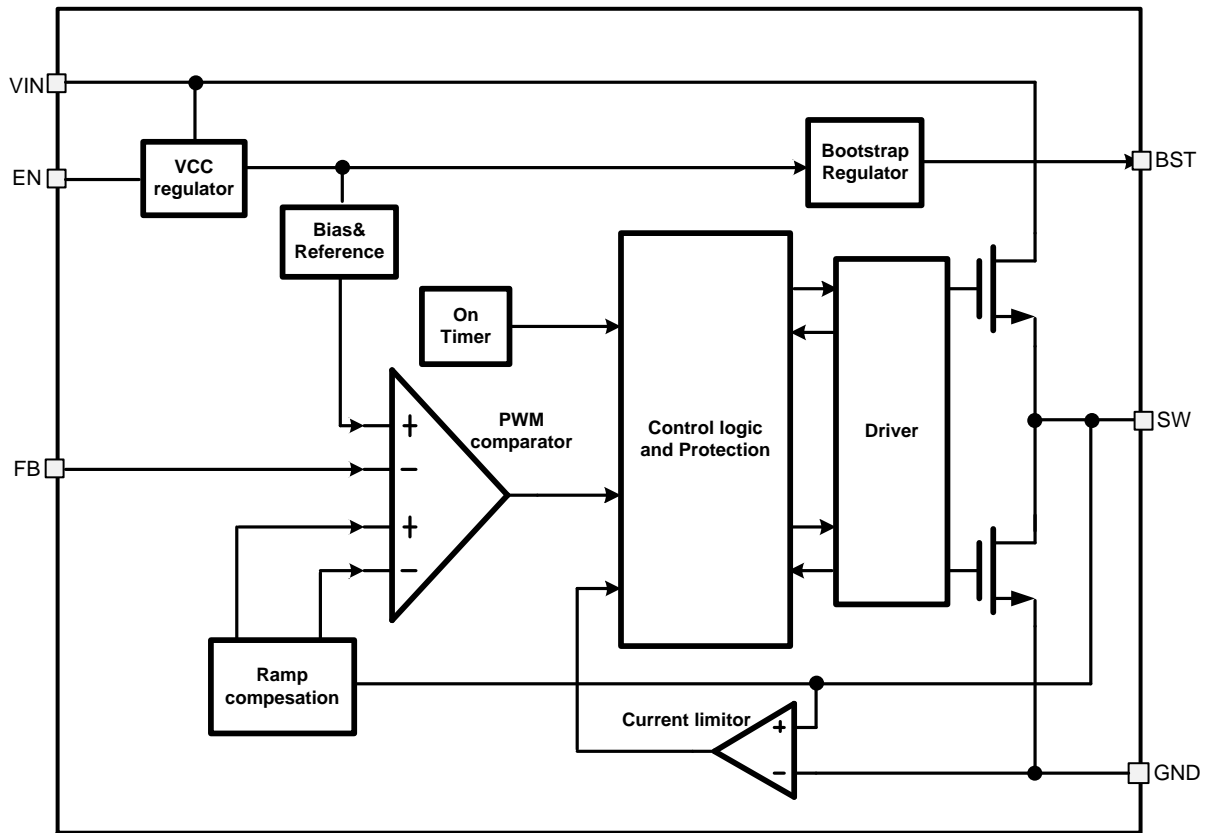
Figure 5. Feedback Voltage vs. Temperature

Figure 6. Quiescent Current vs. Temperature



SCT2220

FUNCTIONAL BLOCK DIAGRAM



SCT2220

Under Voltage Lockout UVLO

The SCT2220 Under Voltage Lock Out (UVLO) default startup threshold is typical 3.9V with VIN rising and shutdown threshold is 3.6V with VIN falling. The more accurate UVLO threshold can be programmed through the precision enable threshold of EN pin.

Enable and Start up

When applying a voltage higher than the EN high threshold (typical 1.2V/rise), the SCT2220 enables all functions and the device starts soft-start phase. The SCT2220 has the built in 2.5ms soft-start time to prevent the output overshoot and inrush current. When EN pin is pulled low, the internal SS net will be discharged to ground. Buck operation is disabled when EN voltage falls below its lower threshold (typically 1.1V/fall).

An internal 1.5uA pull up current source connected from internal LDO power rail to EN pin guarantees that floating EN pin automatically enables the device. For the application requiring higher VIN UVLO voltage than the default setup, there is a 5.3uA hysteresis pull up current source on EN pin which configures the VIN UVLO voltage with an off-chip resistor divider R3 and R4, shown in Figure 7. The resistor divider R3 and R4 are calculated by equation (3) and (4).

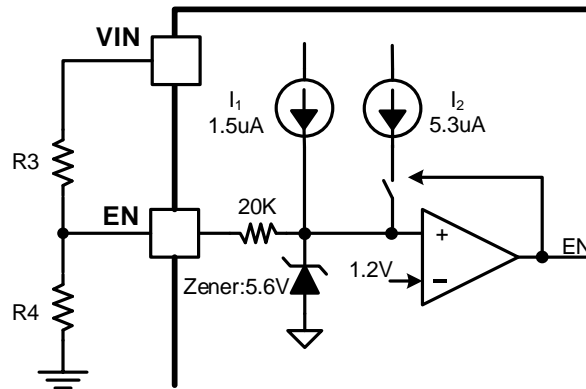


Figure 7. Adjustable VIN UVLO

$$\frac{V_{ENR}}{V_{IN}} = \frac{R4}{R3 + R4} \tag{3}$$

$$V_{ENR} = 1.2V \tag{4}$$

Where:

- Vstart: Vin rise threshold to enable the device
- Vstop: Vin fall threshold to disable the device
- I₁=1.5uA
- I₂=5.3uA
- V_{ENR}=1.2V
- V_{EMF}=1.1V

Over Current Protection (OCP) and Hiccup Mode

In each switching cycle, the inductor current is sensed by monitoring the low-side MOSFET during the OFF period. When the voltage between GND pin and SW pin is lower than the over current threshold voltage, the OCP will be triggered and the controller keeps the OFF state. A new switching cycle will begin only when the measured voltage is higher than limit voltage. If output loading continues to increase, output will dropped below the UVP, and SS pin is discharged such that output is 0V. Then the device will count for 7 cycles of soft-start time for hiccup waiting time and restart normally after 7 cycles soft-start period.

Bootstrap Voltage Regulator

An external bootstrap capacitor between BST and SW pin powers floating high-side power MOSFET gate driver. The bootstrap capacitor voltage is charged from an integrated voltage regulator when high-side power MOSFET is off and low-side power MOSFET is on.

Thermal Shutdown

Once the junction temperature in the SCT2220 exceeds 160°C, the thermal sensing circuit stops converter switching and restarts with the junction temperature falling below 140°C. Thermal shutdown prevents the damage on device during excessive heat and power dissipation condition.

APPLICATION INFORMATION

Typical Application

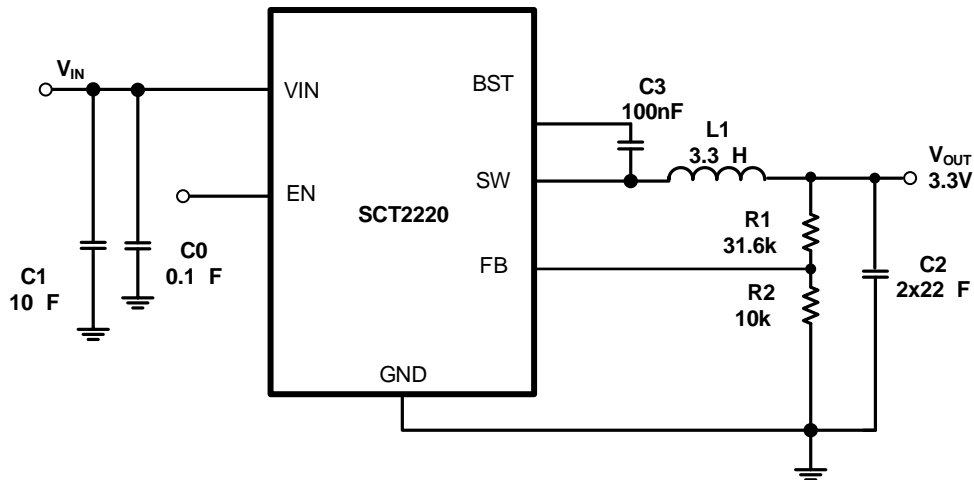


Figure 8. 12V Input, 3.3V/2A Output

Design Parameters

Design Parameters	Example Value
Input Voltage	12V
Output Voltage	3.3V
Output Current	2A
Switching Frequency	750kHz

Input Capacitor Selection

For good input voltage filtering, choose low-ESR ceramic capacitors. A ceramic capacitor 10 μ F for the decoupling capacitor and a 0.1 μ F to be placed as close as possible to the VIN pin of the SCT2220.

Use Equation (5) to calculate the input voltage ripple:

(5)

Where:

- C_{IN} is the input capacitor value
- f_{sw} is the converter switching frequency
- I_{OUT} is the maximum load current

Due to the inductor current ripple, the input voltage changes if there is parasitic inductance and resistance between the power supply and the VIN pin. It is recommended to have enough input capacitance to make the input voltage ripple less than 100mV. Generally, a 25V/10 μ F input ceramic capacitor is recommended for most of

applications. Choose the right capacitor value carefully with considering high-capacitance ceramic capacitors DC bias effect, which has a strong influence on the final effective capacitance.

Inductor Selection

The performance of _____ avior, loop stability, and buck converter efficiency. The inductor value, DC resistance (DCR), and saturation current influences both efficiency and the magnitude of the output voltage ripple. Larger inductance value reduces inductor current ripple and therefore leads to lower output voltage ripple. For a fixed DCR, a larger value inductor yields higher efficiency via reduced RMS and core losses. However, a larger inductor within a given inductor family will generally have a greater series resistance, thereby counteracting this efficiency advantage.

Inductor values can have ±20% or even ±30% tolerance with no current bias. When the inductor current approaches saturation level, its inductance can decrease 20% to 35% from the value at 0-A current depending on how the inductor vendor defines saturation. When selecting an inductor, choose its rated current especially the saturation current larger than its peak current during the operation.

To calculate the current in the worst case, use the maximum input voltage, minimum output voltage, maxim load current and minimum switching frequency of the application, while considering the inductance with -30% tolerance and low-power conversion efficiency.

For a buck converter, calculate the inductor minimum value as shown in equation (6).

_____ (6)

Where:

K_{IND} is the coefficient of inductor ripple current relative to the maximum output current.

SCT2220

Output Feedback Resistor Divider Selection

The SCT2220 features external programmable output voltage by using a resistor divider network R1 and R2 as shown in the typical application circuit Figure 8. Use equation (8) to calculate the resistor divider values.

$$\text{Output Voltage (V)} = \frac{V_{ref}}{1 + \frac{R1}{R2}} \tag{8}$$

(nF)

Table 2. Recommended Component Selections

Output Voltage (V)	SCT2220	L (μH)	C2 (μF)	C3 (nF)

Application Waveforms

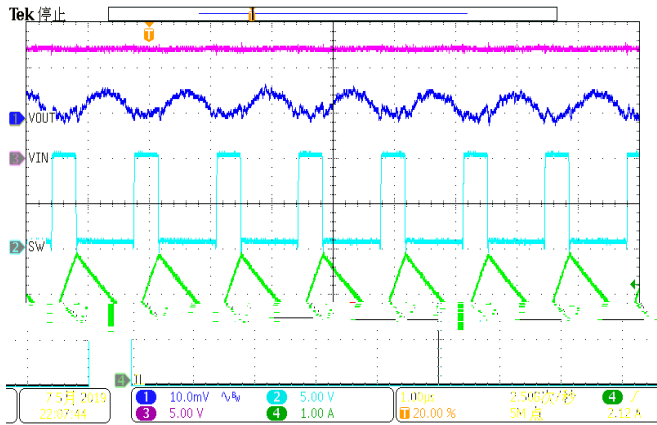


Figure 9. SW node waveform and Output Ripple
VIN=12V, IOUT=2A

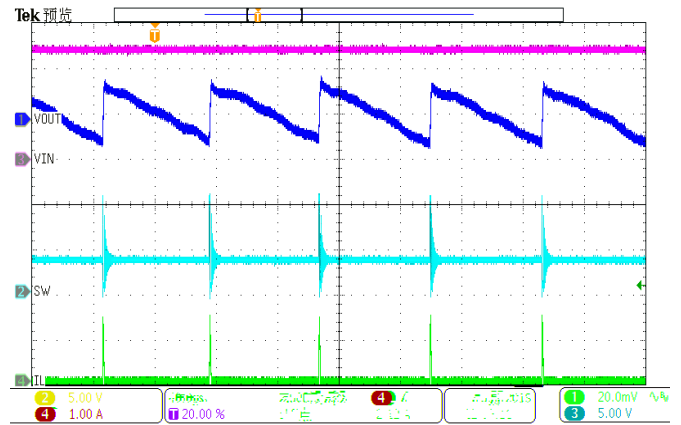


Figure 10. SW node Waveform and Output Ripple
VIN=12V, IOUT=10mA

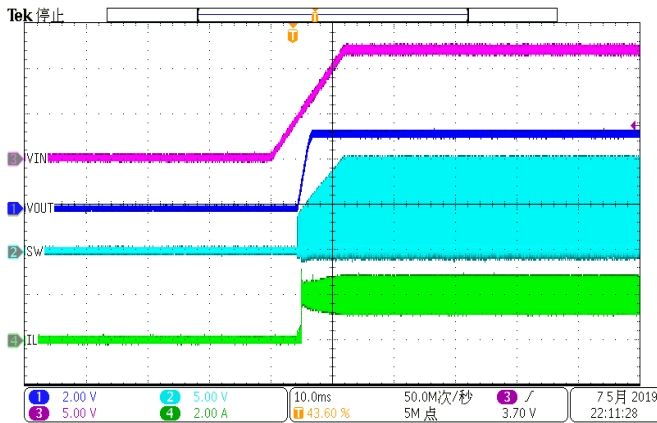


Figure 11. Power Up
VIN=12V, VOUT=3.3V, IOUT=2A

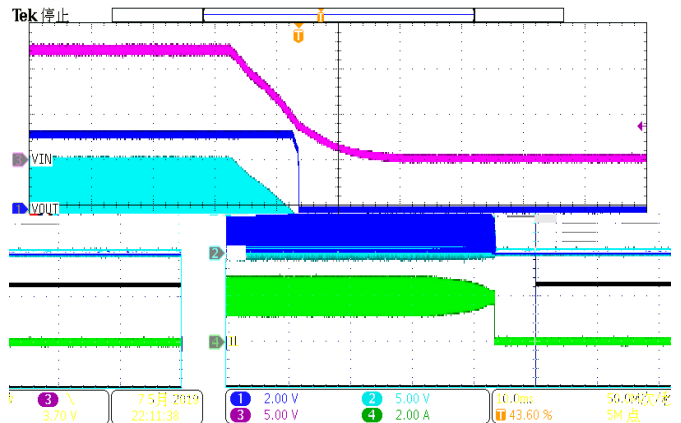


Figure 12. Power Down
VIN=12V, VOUT=3.3V, IOUT=2A

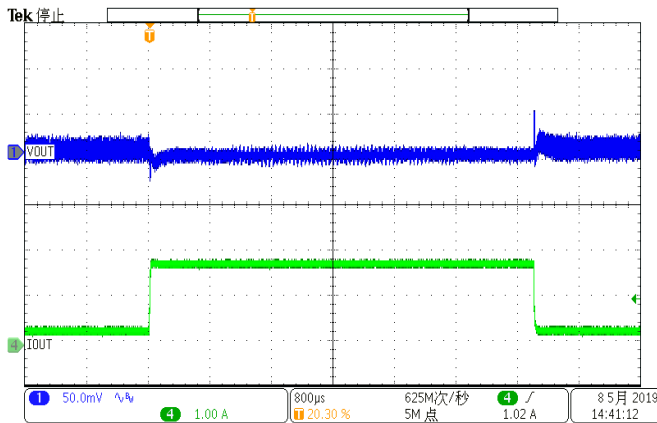


Figure 13. Load Transient
VOUT=3.3V, IOUT=0.2A to 1.8A, SR=250mA/us

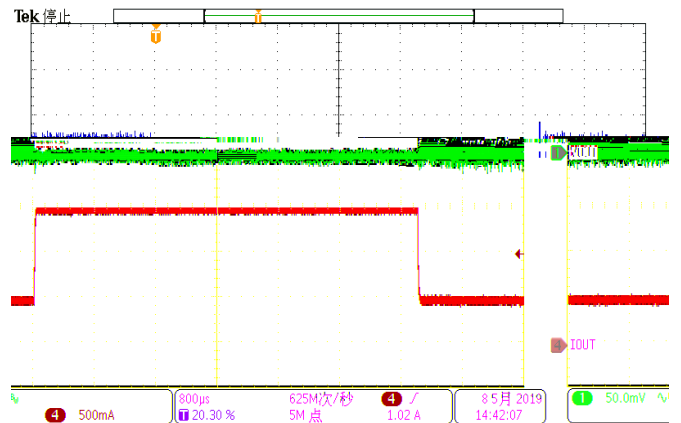


Figure 14. Load Transient
VOUT=3.3V, IOUT=0.5A to 1.5A, SR=250mA/us

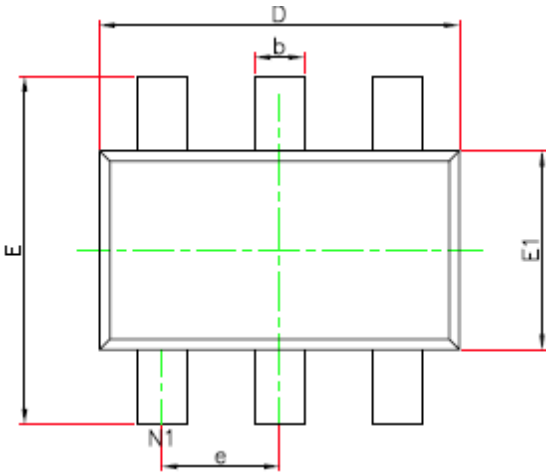
SCT2220

Layout Guideline

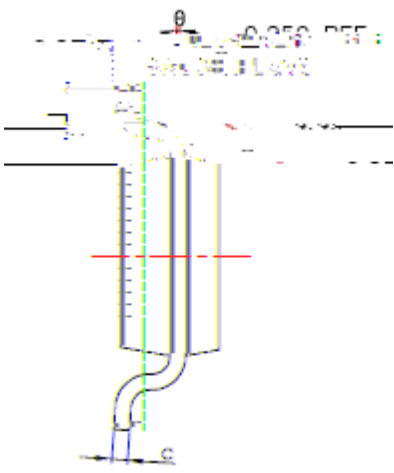
The regulator could suffer from instability and noise problems without carefully layout of PCB. Radiation of high-frequency noise induces EMI, so proper layout of the high-frequency switching path is essential. Minimize the length and area of all traces connected to the SW pin, and always use a ground plane under the switching regulator to minimize coupling. The input capacitor needs to be very close to the VIN pin and GND pin to reduce the input supply ripple. Place the capacitor as close to VIN pin as possible to reduce high frequency ringing voltage on SW pin as well. Figure 15 is the recommended PCB layout of SCT2220.

The layout needs be done with well consideration of the thermal. A large top layer ground plate using multiple thermal vias

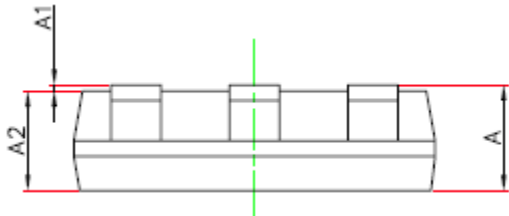
PACKAGE INFORMATION (TSOT23-6)



TSOT23-6 TOP VIEW



TSOT23-6 BOTTOM VIEW



TSOT23-6 SIDE VIEW

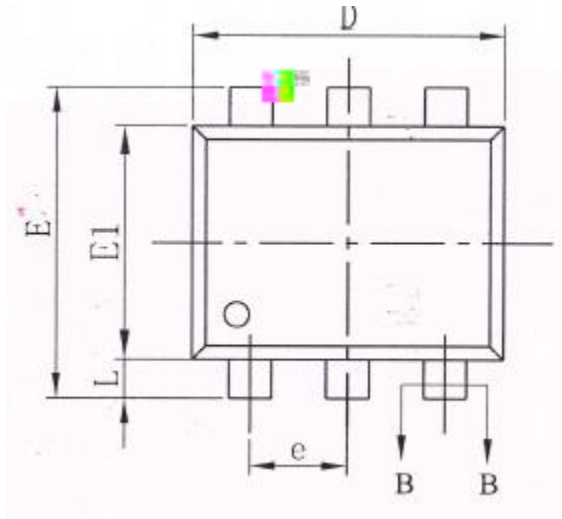
SYMBOL	Unit: Millimeter		
	MIN	TYP	MAX
A	-----		1.10
A1	0.000		0.10
A2	0.70		1.00
D	2.85		2.95
E	2.65		2.95
E1	1.55		1.65
b	0.30		0.50
c	0.08		0.20
e	0.95(BSC)		
L	0.30		0.60
	0		8

NOTE:

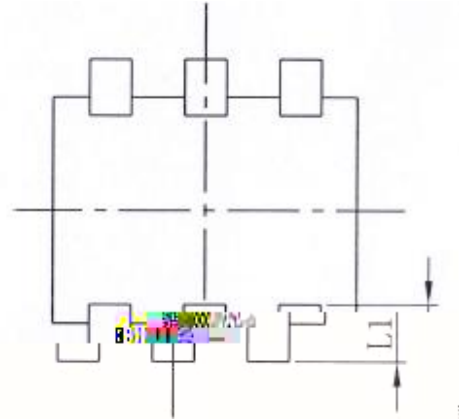
1. Drawing proposed to be made a JEDEC package outline MO-220 variation.
2. Drawing not to scale.
3. All linear dimensions are in millimeters.
4. Thermal pad shall be soldered on the board.
5. Dimensions of exposed pad on bottom of package do not include mold flash.
6. Contact PCB board fabrication for minimum solder mask web tolerances between the pins.

SCT2220

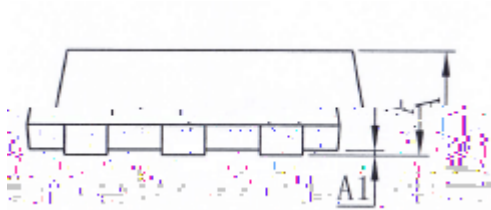
PACKAGE INFORMATION (SOT563)



SOT563 TOP VIEW



SOT563 BOTTOM VIEW



SOT563 SIDE VIEW

NOTE:

7. Drawing proposed to be made a JEDEC package outline MO-220 variation.
8. Drawing not to scale.
9. All linear dimensions are in millimeters.
10. Thermal pad shall be soldered on the board.
11. Dimensions of exposed pad on bottom of package do not include mold flash.
12. Contact PCB board fabrication for minimum solder mask web tolerances between the pins.

SYMBOL	Unit: Millimeter		
	MIN	TYP	MAX
A	0.53		0.6
A1	0.000		0.05
b	0.19		0.27
b1	0.18	0.2	0.23
c	0.11		0.16
c1	0.1	0.11	0.12
D	1.5	1.6	1.7
E	1.5	1.6	1.7
E1	1.1	1.2	1.3
e	0.50BSC		
L	0.1	0.2	0.3
L1	0.2	0.5	0.4

